

Model for the Calculation of Multiply Scattered Fields

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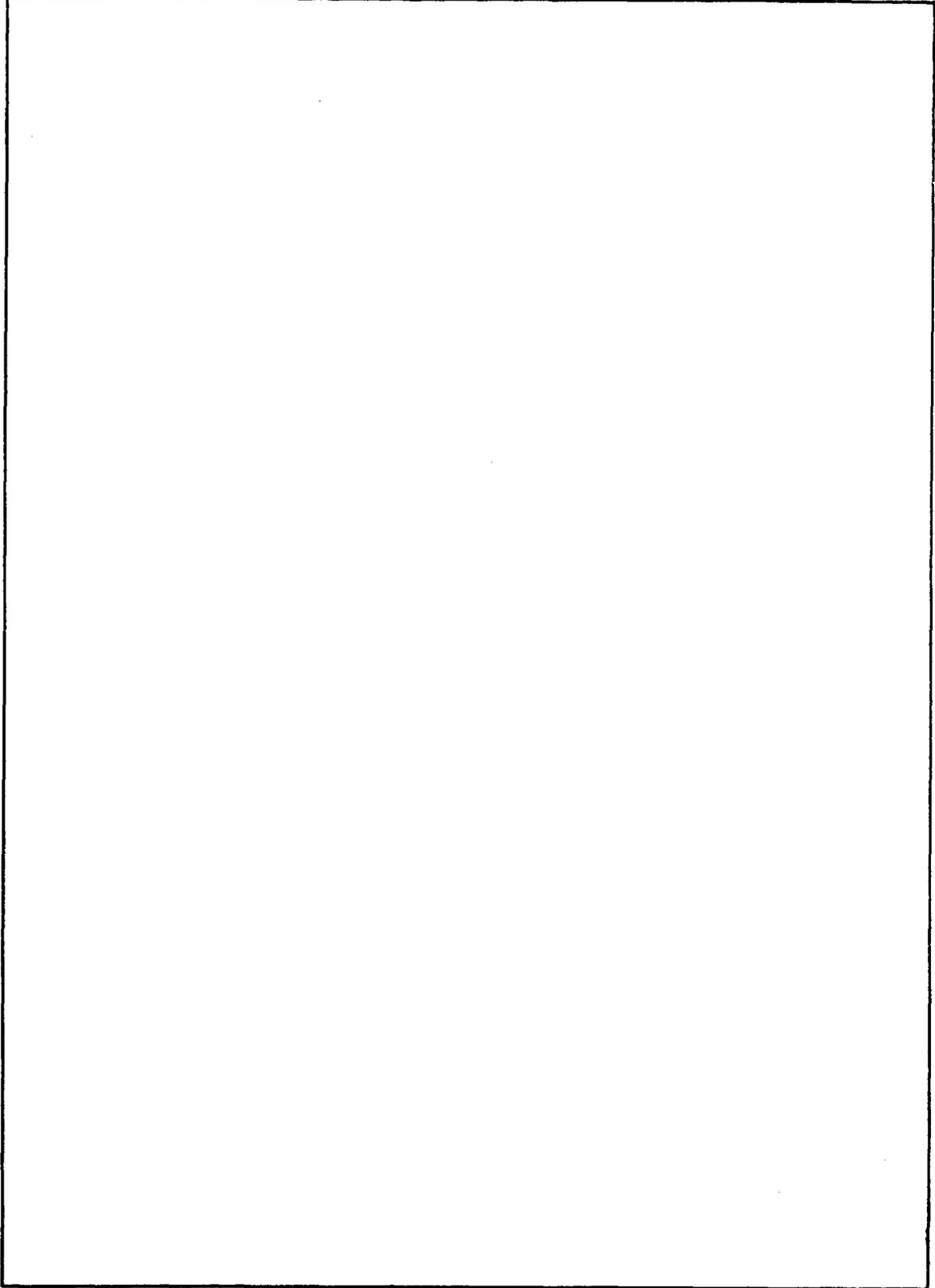
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A MODEL FOR THE CALCULATION OF MULTIPLY SCATTERED FIELDS

INTRODUCTION

This report is a description of a model developed to calculate the field *multiply* scattered by a collection of flat metallic polygonal reflectors. A Fortran listing of the corresponding computer algorithms, programmed for the CDC 3800, is presented as Appendix A. Given a radiating dipole and the vertices of a collection of flat polygonal reflectors, the model produces the field received at any specified sample point (receiver). The field at the receiver consists of the direct ray (transmitter to receiver) combined with indirect rays resulting from multiple scatterings from the polygonal reflectors. The model traces the transmitted radiation through all possible paths in the maze of reflectors, calculates the value of the field after scattering by each reflector encountered, and then sums all contributions to arrive at the total field. In the doppler system the transmitting array elements are energized so as to impart a "virtual velocity" to each transmission, with the result that the signal received at a given point within the beam is shifted in wavelength by an amount proportional to $\sin \theta$, where θ is the angle of the point off the beam center. The model provides for the specification of a single virtual-velocity vector to be associated with each array transmission, separates field-component calculations by wavelength, and produces an energy *spectrum* for each field point sampled.

The program is modular in structure, facilitating the interchange of a number of methods for the calculation of the field scattered from each reflector. At present the program incorporates the standard far-field approximation to the fields scattered by flat plates. The results produced by these calculations may provide only a gross description of system behavior in the near-field and Fresnel regions. More exact scattering formulas that are valid in all three regions are currently under development. Meanwhile the current far-field approach is adequate for program verification and preliminary investigations and will prove valuable in validating the results produced by the model when more sophisticated methods are incorporated.

In the current version of the model, great economy of calculation has been achieved by the use of certain "finite-sum" representations developed by W. B. Gordon [1] for the far-field formulas. The far-field approximation to the field scattered by a flat plate is given by a certain double integral over the reflecting surface [Eq. (1)], and a direct calculation of this integral would involve evaluating the integrand at a large number of points. In the finite-sum representation the evaluation of this integral over a flat, N-sided polygon is reduced to a sum $T_1 + T_2 + \dots + T_N$, where each T_i is a certain complex quantity evaluated at the i th vertex of the polygon.

The discussion to follow first considers the calculation of the field scattered from a single reflector. Then the framework within which such calculations are made, the ray-

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tracing techniques from reflector to reflector, is developed, and the total field spectrum is calculated. Finally, expansions of the model currently underway and areas for future refinement are discussed.

FINITE-SUM REPRESENTATION OF THE STANDARD FAR-FIELD APPROXIMATION TO THE HELMHOLTZ INTEGRAL

Consider a plane wave incident on a flat-plate reflector S of finite dimension. Let \mathbf{f} represent the incident field, with wave number $k = 2\pi/\lambda$.

Let x_1, x_2, x_3 be rectangular coordinates, and let S lie in the x_1x_2 plane. Define R, θ, φ to be spherical coordinates, where

$$x_1 = R \sin \theta \cos \varphi$$

$$x_2 = R \sin \theta \sin \varphi$$

$$x_3 = R \cos \theta.$$

Assume P to be a fixed sample field point in space. If the origin of this coordinate system is located in the interior of S , then \mathbf{g}_P , the standard far-field approximation of the scattered field at P , is given by

$$\mathbf{g}_P = \left[\frac{-ik(1 + \cos \theta) e^{ikR}}{4\pi R} \int_S e^{-ik\mathbf{x} \cdot \boldsymbol{\nu}} dx_1 dx_2 \right] \mathbf{f}_r, \quad (1)$$

where

$$\begin{aligned} \mathbf{f}_r &= \text{incident field } \mathbf{f} \text{ reflected from and evaluated at } S, \\ \mathbf{x} &= [x_1, x_2], \end{aligned}$$

and (2)

$$\boldsymbol{\nu} = [\nu_1, \nu_2] = [\sin \theta \cos \varphi, \sin \theta \sin \varphi].$$

(See Refs. 2 and 3.)

For the finite-sum representation, let S be a flat polygon with N vertices $\mathbf{a}_1, \dots, \mathbf{a}_N$, where each vertex \mathbf{a}_i is a 2-vector in the x_1x_2 plane. Set $\mathbf{a}_{N+1} = \mathbf{a}_1$, and define $\Delta \mathbf{a}_n = \mathbf{a}_{n+1} - \mathbf{a}_n$, $1 \leq n \leq N$. Define $\boldsymbol{\eta}$ to be the unit normal to the plane of S , pointing into the half space containing P . Let the unit normal to the incident plane wave be represented by the "ray" $\boldsymbol{\xi}_i$ and the unit normal to the reflected wave by $\boldsymbol{\xi}_r$, which is given by

$$\boldsymbol{\xi}_r = \boldsymbol{\xi}_i - 2(\boldsymbol{\xi}_i \cdot \boldsymbol{\eta})\boldsymbol{\eta}. \quad (3)$$

Define \mathbf{R} to be the vector from a fixed point in S , such as the centroid, to the field point P , so that $R = |\mathbf{R}|$ represents a "mean" distance from S to P , and let $\boldsymbol{\rho}$ be the projection of \mathbf{R} onto the plane of S :

$$\boldsymbol{\rho} = \mathbf{R} - (\mathbf{R} \cdot \boldsymbol{\eta})\boldsymbol{\eta}. \quad (4)$$

Define \mathbf{w} to be the projection of $[\boldsymbol{\xi} - (\boldsymbol{\rho}/R)]$ onto the plane of S . Since the projections of $\boldsymbol{\xi}_r$ and $\boldsymbol{\xi}_i$ onto the plane of S are equal,

$$\mathbf{w} = \boldsymbol{\xi}_r - (\boldsymbol{\xi}_r \cdot \boldsymbol{\eta})\boldsymbol{\eta} - \frac{\boldsymbol{\rho}}{R} = \boldsymbol{\xi}_i - (\boldsymbol{\xi}_i \cdot \boldsymbol{\eta})\boldsymbol{\eta} - \frac{\boldsymbol{\rho}}{R}. \quad (5)$$

Then Gordon has shown in Ref. 1 that the standard far-field approximation to the Helmholtz integral, given in Eq. (1), can be reduced to the following forms involving no integrations at all:

(a) for $\mathbf{w} \neq 0$,

$$\mathbf{g}_p = \frac{\left(\frac{\mathbf{R}}{R} - \boldsymbol{\xi}_r\right) \cdot \boldsymbol{\eta}}{\mathbf{w} \cdot \mathbf{w}} \frac{e^{ikR}}{4\pi R} (T_1 + \dots + T_N) \mathbf{f}_r, \quad (6)$$

where, for $1 \leq n \leq N$,

$$T_n = [(\mathbf{w} \times \boldsymbol{\eta}) \cdot \Delta \mathbf{a}_n] \frac{\sin \left[\frac{k}{2} \mathbf{w} \cdot \Delta \mathbf{a}_n \right]}{\left[\frac{k}{2} \mathbf{w} \cdot \Delta \mathbf{a}_n \right]} \exp \left[\frac{ik}{2} \mathbf{w} \cdot (\mathbf{a}_n + \mathbf{a}_{n+1}) \right], \quad (7)$$

and

(b) for $\mathbf{w} = 0$,

$$\mathbf{g}_p = -ik \left[\frac{\mathbf{R}}{R} \cdot \boldsymbol{\eta} \right] \frac{e^{ikR}}{2\pi R} A \mathbf{f}_p, \quad (8)$$

where A is the area of S . Note that geometrically $\mathbf{w} = 0$ implies $\boldsymbol{\xi}_r = \mathbf{R}/R$, which means that the specular reflection of the incident wave is in the direction of the field point P .

The following sections illustrate the application of this result in the case of the single and then multiple reflector configurations.

MODEL DEFINITIONS AND PRELIMINARY CALCULATIONS

Let

$\mathbf{E}_T = [E_{Tx}, E_{Ty}, E_{Tz}] =$ transmitted radiation
(indicating polarization),

$\lambda =$ transmitting wavelength,

and

$\mathbf{R}_T = [X_T, Y_T, Z_T] =$ coordinates of transmitter,
assuming a right-hand coordinate system.

Define the reflector S by specifying

$N =$ number of vertices of the reflecting polygon

and

$\mathbf{V}_n = [V_{nx}, V_{ny}, V_{nz}] =$ coordinates of the n th vertex, $1 \leq n \leq N$.

The vertices are taken in a positive direction around the reflector, which is defined to be in counterclockwise order as viewed from the transmitter.

Let \mathbf{VC} be the centroid of the reflecting plate S ,

$$\mathbf{VC} = \frac{\mathbf{V}_1 + \mathbf{V}_2 + \dots + \mathbf{V}_N}{N} \quad (9)$$

and $\boldsymbol{\eta}$ the unit normal to the plate,

$$\boldsymbol{\eta} = \frac{\pm (\mathbf{V}_1 - \mathbf{VC}) \times (\mathbf{V}_2 - \mathbf{VC})}{|\mathbf{V}_1 - \mathbf{VC}| |\mathbf{V}_2 - \mathbf{VC}|}, \quad (10)$$

where the sign of $\boldsymbol{\eta}$ is chosen so that $\boldsymbol{\eta}$ points into the half space containing the transmitter,

$$(\mathbf{R}_T - \mathbf{VC}) \cdot \boldsymbol{\eta} \geq 0.$$

The unit normal to the incident plane wave, $\boldsymbol{\xi}_i$, is given by

$$\boldsymbol{\xi}_i = \frac{\mathbf{VC} - \mathbf{R}_T}{|\mathbf{VC} - \mathbf{R}_T|}, \quad (11)$$

and the unit normal to the reflected wave, $\boldsymbol{\xi}_r$, which is obtained by reflecting $\boldsymbol{\xi}_i$ about $\boldsymbol{\eta}$, is given by

$$\xi_r = \xi_i - 2(\xi_i \cdot \eta)\eta. \tag{12}$$

Fig. 1 illustrates a typical single-reflector situation.

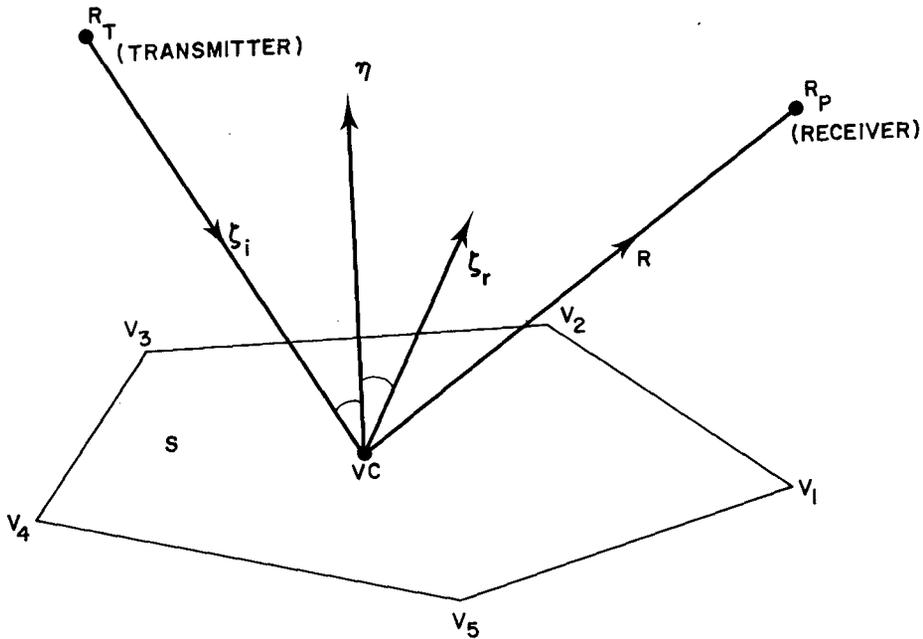


Fig. 1 — Geometry of a typical single-reflector case

The area A of the reflecting polygon can be calculated by dividing the polygon into N triangles of area Δ_n , each with a vertex at VC (Fig. 2),

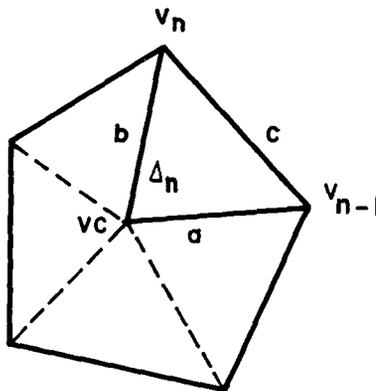


Fig. 2 — Division of the reflecting polygon for the area calculation

so that

$$\Delta_n = \frac{1}{4} [2(a^2b^2 + b^2c^2 + c^2a^2) - (a^4 + b^4 + c^4)]^{1/2}, \quad (13)$$

where

$$a = |\mathbf{V}_{n-1} - \mathbf{VC}|, \quad b = |\mathbf{V}_n - \mathbf{VC}|, \quad c = |\mathbf{V}_n - \mathbf{V}_{n-1}|. \quad (14)$$

Thus it follows that

$$A = \sum_{n=1}^N \Delta_n. \quad (15)$$

FIELD CALCULATION FOR A SINGLE REFLECTOR

Let $\mathbf{R}_p = [X_p, Y_p, Z_p]$ represent the position of the field sample point P at which the total field is to be evaluated. Assuming a dipole radiator, the direct-ray contribution to the total field at P is given by

$$\mathbf{E}_D(\mathbf{R}_p) = \left\{ \mathbf{E}_T - \left[\frac{\mathbf{E}_T \cdot (\mathbf{R}_p - \mathbf{R}_T)}{|\mathbf{R}_p - \mathbf{R}_T|^2} \right] (\mathbf{R}_p - \mathbf{R}_T) \right\} \frac{\exp(ik|\mathbf{R}_p - \mathbf{R}_T|)}{|\mathbf{R}_p - \mathbf{R}_T|}. \quad (16)$$

The scattered-ray component of the total field consists of the radiation reflected from the surface S and scattered in the direction of P . The direct field at the centroid \mathbf{VC} of plate S is

$$\mathbf{F} = \frac{\mathbf{E}_T - \left[\frac{\mathbf{E}_T \cdot (\mathbf{VC} - \mathbf{R}_T)}{|\mathbf{VC} - \mathbf{R}_T|^2} \right] (\mathbf{VC} - \mathbf{R}_T)}{|\mathbf{VC} - \mathbf{R}_T|}, \quad (17)$$

which, on reflection from S , becomes

$$\mathbf{F}_r = [2(\mathbf{F} \cdot \boldsymbol{\eta})\boldsymbol{\eta} - \mathbf{F}] \exp[ik|\mathbf{VC} - \mathbf{R}_T|], \quad (18)$$

where the first factor represents the optical reflection of the incident field and the second factor contributes the proper phase.

The next step is to apply the finite-sum representation of the far-field approximation to the Helmholtz integral, as stated in Eqs. (6) through (8), to the reflected field \mathbf{F}_r to obtain the scattered field at \mathbf{R}_p . However, if the source \mathbf{R}_T of the incident field and the location \mathbf{R}_p of the sample point are on opposite sides of the reflector, there is no scattered-field component at \mathbf{R}_p , since shadowing is not currently considered in this model. In this case, no further calculations are made. Assume therefore that \mathbf{R}_p and \mathbf{R}_T are on the same

side of the plane containing the reflector. As in the second section, define \mathbf{R} to be the vector from the centroid of S to the sample field point, so that

$$\mathbf{R} = \mathbf{R}_P - \mathbf{VC}, \tag{19}$$

and

$$R = |\mathbf{R}|. \tag{20}$$

The projection of \mathbf{R} onto the plane of S is given by

$$\boldsymbol{\rho} = \mathbf{R} - (\mathbf{R} \cdot \boldsymbol{\eta})\boldsymbol{\eta}, \tag{21}$$

and the projection of $[\boldsymbol{\xi} - (\boldsymbol{\rho}/R)]$ onto the same plane is given by

$$\mathbf{w} = \boldsymbol{\xi}_r - (\boldsymbol{\xi}_r \cdot \boldsymbol{\eta})\boldsymbol{\eta} - \frac{\boldsymbol{\rho}}{R} = \boldsymbol{\xi}_i - (\boldsymbol{\xi}_i \cdot \boldsymbol{\eta})\boldsymbol{\eta} - \frac{\boldsymbol{\rho}}{R}, \tag{22}$$

since the projections of $\boldsymbol{\xi}_r$ and $\boldsymbol{\xi}_i$ onto S are equal. Figure 3 illustrates the geometry involved.

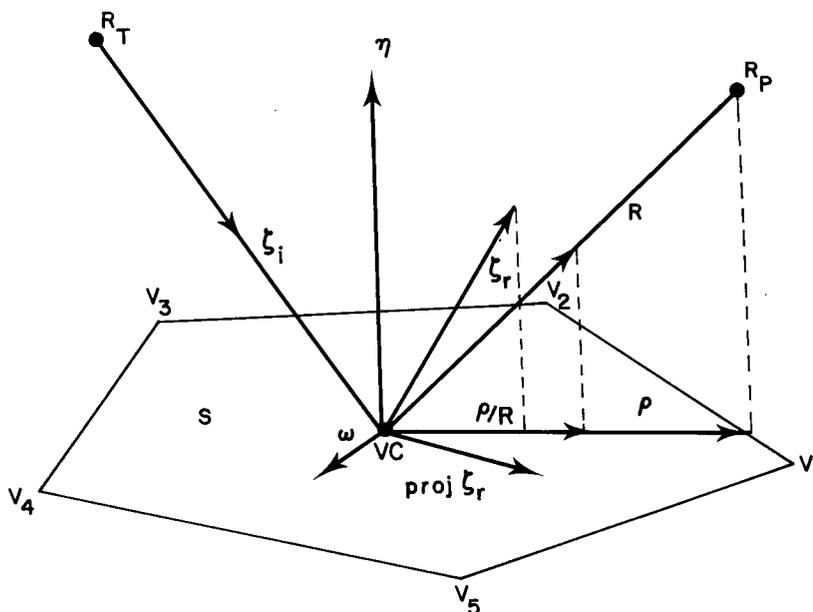


Fig. 3 — Geometry for the calculation of \mathbf{w}

In the case where $\mathbf{w} = 0$, indicating that the specular reflection of the incident wave is in the direction of the field point P , so that $\xi_r = \mathbf{R}/R$, the scattered-field component is given by Eq. (8), yielding

$$\mathbf{E}_{SC}(\mathbf{R}_P) = \left\{ -ik \left[\frac{\mathbf{R}}{R} \cdot \boldsymbol{\eta} \right] \frac{e^{ikR}}{2\pi R} A \right\} \mathbf{F}_r. \quad (23)$$

If $\mathbf{w} \neq 0$, then by Eqs. (6) and (7), the scattered field component is

$$\mathbf{E}_{SC}(\mathbf{R}_P) = \left[\frac{\left(\frac{\mathbf{R}}{R} \quad \xi_r \right) \cdot \boldsymbol{\eta}}{\mathbf{w} \cdot \mathbf{w}} \frac{e^{ikR}}{4\pi R} \sum_{n=1}^N T_n \right] \mathbf{F}_r, \quad (24)$$

where for each vertex n , $1 \leq n \leq N$,

$$T_n = [(\mathbf{w} \times \boldsymbol{\eta}) \cdot \Delta \mathbf{V}_n] \frac{\sin \left[\frac{k}{2} \mathbf{w} \cdot \Delta \mathbf{V}_n \right]}{\left[\frac{k}{2} \mathbf{w} \cdot \Delta \mathbf{V}_n \right]} \exp \left[\frac{ik}{2} \mathbf{w} \cdot (\mathbf{V}_n + \mathbf{V}_{n+1}) \right] \quad (25)$$

in which

$$\Delta \mathbf{V}_n = \mathbf{V}_{n+1} - \mathbf{V}_n, \quad (26)$$

$$\mathbf{V}_{N+1} = \mathbf{V}_1.$$

The total field at P is then the sum of the direct and scattered fields evaluated at \mathbf{R}_P :

$$\mathbf{E}(\mathbf{R}_P) = \mathbf{E}_D(\mathbf{R}_P) + \mathbf{E}_{SC}(\mathbf{R}_P). \quad (27)$$

FIELD CALCULATION FOR MULTIPLE REFLECTORS

When two or more reflectors are involved, it becomes necessary to keep track of rays bouncing back and forth between them while calculating the scattering taking place at each surface. The contributions from all these multiple scatterings, along with the direct ray, then constitute the total field seen at any given sample point. This section describes the approach used within the model to systematically exhaust all possible ray paths among a given set of reflectors and arrive at the total scattered field. The preliminary calculations initially parallel those of the single-reflector case, with the addition of indices to keep track of individual reflectors.

Consider a configuration with M flat polygonal reflectors S_1, S_2, \dots, S_M , where $2 \leq M < \infty$. As in the case of a single reflector, let

$$\mathbf{E}_T = (E_{Tx}, E_{Ty}, E_{Tz})$$

be the transmitted radiation, and let

$$\mathbf{R}_T = [X_T, Y_T, Z_T]$$

the position of the transmitter. As a logical extension of the earlier case, define N_i to be the number of vertices of the i th reflector S_i , $1 \leq i \leq M$ and define

$$\mathbf{V}_{ik} = [V_{ikx}, V_{iky}, V_{ikz}]$$

to be the coordinates of the k th vertex of S_i , $1 \leq k \leq N_i$. Then

$$\mathbf{VC}_i = \frac{(\mathbf{V}_{i1} + \mathbf{V}_{i2} + \dots + \mathbf{V}_{iN_i})}{N_i} \quad (28)$$

is the centroid of the i th reflecting plate, and

$$\boldsymbol{\eta}_i = \pm \frac{(\mathbf{V}_{i1} - \mathbf{VC}_i) \times (\mathbf{V}_{i2} - \mathbf{VC}_i)}{|\mathbf{V}_{i1} - \mathbf{VC}_i| |\mathbf{V}_{i2} - \mathbf{VC}_i|}, \quad (29)$$

where again the sign of $\boldsymbol{\eta}_i$ is taken such that

$$(\mathbf{R}_T - \mathbf{VC}_i) \cdot \boldsymbol{\eta}_i \geq 0,$$

is the unit normal to S_i .

Let A_i represent the area of the i th plate, so that

$$A_i = \sum_{n=1}^{N_i} \Delta_{in}, \quad (30)$$

where

$$\Delta_{in} = \frac{[2(a^2b^2 + b^2c^2 + c^2a^2) - (a^4 + b^4 + c^4)]^{1/2}}{4},$$

$$a = |\mathbf{V}_{i(n-1)} - \mathbf{VC}_i|,$$

$$b = |\mathbf{V}_{in} - \mathbf{VC}_i|,$$

and

$$c = |\mathbf{V}_{in} - \mathbf{V}_{i(n-1)}|.$$

Define \mathbf{R}_{ij} to be the unit direction vector from the centroid of reflector S_i to the centroid of reflector S_j . Thus

$$\mathbf{R}_{ij} = \frac{\mathbf{VC}_j - \mathbf{VC}_i}{|\mathbf{VC}_j - \mathbf{VC}_i|} \quad (31)$$

for $0 \leq i \leq M$, $1 \leq j \leq M$, $i \neq j$.

Note that $\mathbf{R}_{0j} = (\mathbf{VC}_j - \mathbf{R}_T)/|\mathbf{VC}_j - \mathbf{R}_T|$ represents the unit vector from the transmitter \mathbf{R}_T (indicated by subscript 0) to the centroid of S_j , which is the quantity defined as ξ_i in Eq. (11) for the single-reflector case. Thus each \mathbf{R}_{ij} gives the direction of radiation originating from reflector S_i (radiation actually being scattered from S_i when $i \neq 0$) incident on reflector S_j .

The preliminary quantities defined and calculated so far have followed closely those of the single-reflector case, as would be expected. In review, associated with each reflector S_i , $1 \leq i \leq M$, are the centroid \mathbf{VC}_i , the normal $\boldsymbol{\eta}_i$ to the plane containing S_i , the area A_i , and a set of unit direction vectors \mathbf{R}_{ij} , $1 \leq j \leq M$, $j \neq i$, which point from S_i to each of the other reflectors S_j . Note that the indices i and j as used here denote reflectors; when used together, such as in \mathbf{R}_{ij} , index i refers to the "source" reflecting plate and index j to the scattering plate. The index k serves as a reflector-vertex index, where $1 \leq k \leq N_j$, $1 \leq j \leq M$.

With the preliminary calculations completed, the actual scattered-field computations for field point P begin with the selection of an initial reflector S_i and the corresponding computation of the direct field at \mathbf{VC}_i due to transmission from \mathbf{R}_T , minus the phase factor, given by

$$\mathbf{F}_0(\mathbf{VC}_i) = \frac{\mathbf{E}_T - \left[\frac{\mathbf{E}_T \cdot (\mathbf{VC}_i - \mathbf{R}_T)}{|\mathbf{VC}_i - \mathbf{R}_T|^2} \right] (\mathbf{VC}_i - \mathbf{R}_T)}{|\mathbf{VC}_i - \mathbf{R}_T|} \quad (32)$$

If the magnitude of \mathbf{F}_0 falls below the threshold,

$$|\mathbf{F}_0(\mathbf{VC}_i)| < \frac{\epsilon A_i |E_D(\mathbf{R}_P)|}{A_1 + A_2 + \dots + A_M}, \quad (33)$$

where the weighing factor ϵ is selected by the user and $E_D(\mathbf{R}_P)$ is the direct field at P , then the field contribution of \mathbf{F}_0 scattered from S_i to the total field at P is considered negligible, and S_i is replaced as an initial reflector. However, if $\mathbf{F}_0(\mathbf{VC}_i)$ exceeds the threshold in Eq. (33), then $|\mathbf{F}_0(\mathbf{VC}_i)|$ is reflected from S_i , with the appropriate change in phase, so that the reflected $\mathbf{F}_{0i}(\mathbf{VC}_i)$, evaluated at the centroid \mathbf{VC}_i is given by

$$\mathbf{F}_{0i}(\mathbf{VC}_i) = [2\mathbf{F}_0(\mathbf{VC}_i) \cdot \boldsymbol{\eta}_i] \boldsymbol{\eta}_i - \mathbf{F}_0(\mathbf{VC}_i) \cdot \exp(ik|\mathbf{VC}_i - \mathbf{R}_T|). \quad (34)$$

Note that the subscripts of \mathbf{F} trace the history of the radiation involved; thus \mathbf{F}_{0i} denotes radiation from R_T reflected from S_i .

Let $\mathbf{G}_i(\mathbf{R}_Q)$ represent the reflected field \mathbf{F}_{0i} scattered by S_i , evaluated at the arbitrary point Q . By Eqs. (6) and (7),

$$\mathbf{G}_i(\mathbf{R}_Q) = \frac{1}{\mathbf{w} \cdot \mathbf{w}} \left[\left(\frac{\mathbf{R}_Q - \mathbf{VC}_i}{|\mathbf{R}_Q - \mathbf{VC}_i|} - \mathbf{R}_{0i} \right) \cdot \boldsymbol{\eta}_i \frac{e^{ik|\mathbf{R}_Q - \mathbf{VC}_i|}}{4\pi|\mathbf{R}_Q - \mathbf{VC}_i|} \sum_{n=1}^N T_n \right] \mathbf{F}_{0i}, \quad (35)$$

where

$$\mathbf{w} = [\mathbf{R}_{0i} - (\mathbf{R}_Q - \mathbf{VC}_i)] - \{ [\mathbf{R}_{0i} - (\mathbf{R}_Q - \mathbf{VC}_i)] \cdot \boldsymbol{\eta}_i \} \boldsymbol{\eta}_i \quad (36)$$

and

$$T_n = [(\mathbf{w} \times \boldsymbol{\eta}_i) \cdot \Delta \mathbf{V}_n] \frac{\sin \left[\frac{k}{2} \mathbf{w} \cdot \Delta \mathbf{V}_n \right]}{\left[\frac{k}{2} \mathbf{w} \cdot \Delta \mathbf{V}_n \right]} \exp \left[\frac{ik}{2} \mathbf{w} \cdot (\mathbf{V}_n + \mathbf{V}_{n+1}) \right], \quad (37)$$

with

$$\Delta \mathbf{V}_n = \mathbf{V}_{i(n+1)} - \mathbf{V}_{i(n)}$$

and

$$\mathbf{V}_{i(N_i+1)} = \mathbf{V}_{i1} \quad (38)$$

for $1 \leq n \leq N_i$. Thus in particular the value of the field at P resulting from the scattering of \mathbf{F}_{0i} from S_i is given by $\mathbf{G}_i(\mathbf{R}_P)$, and the value of the scattered field at the centroid of each of the reflecting plates other than S_i by

$$\mathbf{G}_i(\mathbf{VC}_j), \quad 1 \leq j \leq M, \quad j \neq i.$$

Consider a second reflector $S_j, j \neq i$. If, as in Eq. (33),

$$|\mathbf{G}_i(\mathbf{VC}_j)| < \frac{\epsilon A_j |\mathbf{E}_D(\mathbf{R}_P)|}{A_1 + A_2 + \dots + A_M}, \quad (39)$$

then the contribution of any further scattering of this ray path is taken to be insignificant, and the path (R_T to S_i to S_j) is abandoned. Consequently a new second reflector $S_k, k \neq i, k \neq j$, is selected, and the Eq. (39) threshold check is repeated.

Assume, however, that the condition of Eq. (39) is not met, that is the value of the scattered field at the centroid of reflector S_j exceeds the specified threshold. The path

(R_T to S_i to S_j) is then pursued. By Eq. (34) the field due to the reflection of F_{0i} at the centroid of S_j is given by

$$F_{0ij} = [(2\mathbf{G}_i \cdot \boldsymbol{\eta}_j)\boldsymbol{\eta}_j - \mathbf{G}_i]. \quad (40)$$

Let \mathbf{G}_{ij} represent the scattering of F_{0ij} from S_j , as given by Eq. (35). Then $G_{ij}(R_P)$ is the field contribution of the ray path (R_T to S_i to S_j to R_P) to the total field at P . Likewise, $\mathbf{G}_{ij}(VC_k)$ is the value of this scattered field at the centroid of S_k for all $1 \leq k \leq M$, $k \neq j$. Third-level reflection, where a reflector S_k , $k \neq j$, is selected, is treated in the same way as the second-level reflection just discussed. Additional levels of reflection are considered along each path until the path is terminated by either failing to meet the threshold at a given reflector and finding that all other reflectors for this path at this level have been considered or exceeding a predetermined maximum on the number of reflection levels allowed. A limit is set on the number of "bounces" between reflectors a field ray may take to facilitate bookkeeping within the computer implementation of the model. In practice the bounce or reflection-level limit is set high enough that paths are invariably terminated by failing to meet the threshold.

It is evident then that when one path is terminated at a particular reflector on a given level of reflection, a new reflector is to be selected at the same reflection level, subject only to the conditions that the chosen reflector is different from that of the preceding level (insuring a distinct bounce) and has not been previously considered for the current path base (as defined up through the preceding level). If all reflectors have been exhausted on this level, attention is returned to the previous level, a new path is defined by selecting a new reflector at this now current level, and the whole process moves forward again. Thus all paths are exhausted when, in the process of stepping down level by level in an attempt to define a new path, all reflectors are found to have been already considered on each level, down through the initial level of reflection.

An example may clarify the steps outlined previously in the generation of all possible ray paths. Consider a configuration consisting of three reflectors S_i , $1 \leq i \leq 3$, a transmitter at R_T , and field sample point at R_P . For purposes of illustration assume that no reflector shadows another, impose a bounce limit of three on any single ray path, and ignore any threshold restrictions. Figure 4 shows all possible ray paths for this situation. These paths are generated, as indicated previously, by starting at the initial level or first tier of reflection in the diagram, directly above R_T , with the lowest indexed available reflector S_1 . Thus path 1 is simply (R_T to S_1 to R_P). Note that in Fig. 4 the last leg of each path (from the reflector to field point R_P) is indicated by a dashed rather than solid line for easier identification. For the next path the lowest indexed available (different from S_1 and not previously selected) reflector in the second tier, S_2 , is chosen, so that path 2 is (R_T to S_1 to S_2 to R_P). Moving along to the third level, path 3 becomes (R_T to S_1 to S_2 to S_1 to R_P) through the selection of S_1 on tier 3. Since a bounce limit of three has been adopted for this case (illustrated in the figure by the existence of only three reflection tiers), path 4 requires a step down in the

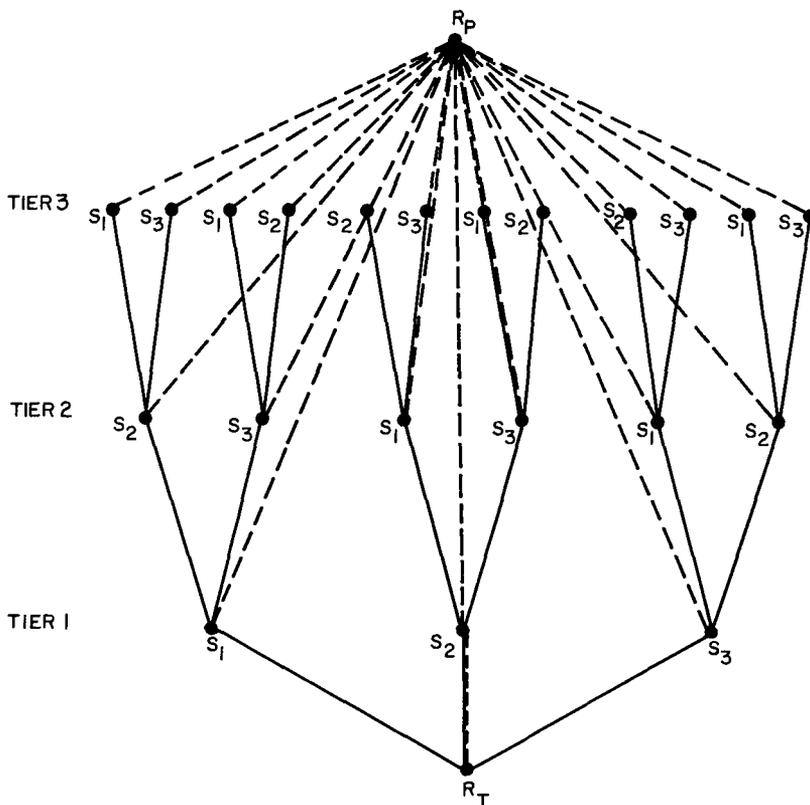


Fig. 4 — Ray paths

diagram back to the second tier, to S_2 , and a reevaluation of the reflectors still available at the third level. Reflector S_3 on level 3 is the only available reflector not yet used on the current path base (R_T to S_1 to S_2), so the selection of S_3 on level 3 defines path 4 as (R_T to S_1 to S_2 to S_3 to R_P). Note reflector S_2 is ineligible for consideration at this level for path base (R_T to S_1 to S_2) since it would not constitute a distinct bounce. Again attention is returned to level 2 at S_2 since the third level allows no further reflections, but now all available third-tier reflectors for path base (R_T to S_1 to S_2) have been selected in previous paths. Thus a second step back, to S_1 on level one, is made. From the vantage point of S_1 , with path base (R_T to S_1), S_3 remains as the only available reflector on level two, so path 5 becomes (R_T to S_1 to S_3 to R_P). Paths 6, as (R_T to S_1 to S_3 to S_1 to R_P), and 7, as (R_T to S_1 to S_3 to S_2 to R_P), follow in the same way. Stepping down now to S_1 on level 1 reveals that all level 2 alternatives for path base (R_T to S_1) have been used (namely S_2 and S_3), so the next lowest indexed reflector on level one, reflector S_2 , is selected. Thus path 8 is (R_T to S_2 to R_P), and the whole process described previously for S_1 is repeated. When attention is returned

to the initial level for the third time, the remaining reflector S_3 is used, the preceding procedure is reiterated, and the generation of all possible ray paths for this configuration is completed.

Note that Fig. 4 is composed of three vertical sections, one for each reflector, as enumerated in the first tier. Using the diagram, the paths described can be traced by starting from the leftmost section and, within that section, moving along the leftmost solid lines from tier to tier, working toward the right. At each reflector in a given tier, a dotted line directly to R_P represents the completion of a path. Table 1 lists all possible paths and corresponding field contributions for this example in the order of their generation by the previously outlined steps.

Table 1
Paths and Corresponding Field Contributions

Path Number	Path	Field Contributions at R_P
0	R_T to R_P	$E_D(R_P)$
1	R_T to S_1 to R_P	$G_1(R_P)$
2	R_T to S_1 to S_2 to R_P	$G_{12}(R_P)$
3	R_T to S_1 to S_2 to S_1 to R_T	$G_{121}(R_P)$
4	R_T to S_1 to S_2 to S_3 to R_P	$G_{123}(R_P)$
5	R_T to S_1 to S_3 to R_P	$G_{13}(R_P)$
6	R_T to S_1 to S_3 to S_1 to R_P	$G_{131}(R_P)$
7	R_T to S_1 to S_3 to S_2 to R_P	$G_{132}(R_P)$
8	R_T to S_2 to R_P	$G_2(R_P)$
9	R_T to S_2 to S_1 to R_P	$G_{21}(R_P)$
10	R_T to S_2 to S_1 to S_2 to R_P	$G_{212}(R_P)$
11	R_T to S_2 to S_1 to S_3 to R_P	$G_{213}(R_P)$
12	R_T to S_2 to S_3 to R_P	$G_{23}(R_P)$
13	R_T to S_2 to S_3 to S_1 to R_P	$G_{231}(R_P)$
14	R_T to S_2 to S_3 to S_2 to R_P	$G_{232}(R_P)$
15	R_T to S_3 to R_P	$G_3(R_P)$
16	R_T to S_3 to S_1 to R_P	$G_{31}(R_P)$
17	R_T to S_3 to S_1 to S_2 to R_P	$G_{312}(R_P)$
18	R_T to S_3 to S_1 to S_3 to R_P	$G_{313}(R_P)$
19	R_T to S_3 to S_2 to R_P	$G_{32}(R_P)$
20	R_T to S_3 to S_2 to S_1 to R_P	$G_{321}(R_P)$
21	R_T to S_3 to S_2 to S_3 to R_P	$G_{323}(R_P)$

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The general expression for the total field at the sample point P is thus

$$\begin{aligned} \mathbf{E}_T(\mathbf{R}_P) = \mathbf{E}_D(\mathbf{R}_P) + \sum_{i=1}^M \left[\mathbf{G}_i(\mathbf{R}_P) + \sum_{\substack{j=1 \\ j \neq i}}^M \left[\mathbf{G}_{ij}(\mathbf{R}_P) + \sum_{\substack{k=1 \\ k \neq j}}^M \left[\mathbf{G}(\mathbf{R}_P)_{ijk} \right. \right. \right. \\ \left. \left. \left. + \dots + \sum_{\substack{n=1 \\ n \neq n-1}}^M \left[\mathbf{G}(\mathbf{R}_P)_{ijk \dots n} \dots \right] \dots \right] \right] \right], \end{aligned} \quad (41)$$

where $\mathbf{E}_D(\mathbf{R}_P)$ is the direct field at P , $\mathbf{G}_{ijk}(\mathbf{R}_P)$ is the field contribution at P due to path (R_T to S_i to S_j to S_k to R_P), and n is the maximum reflection level.

DOPPLER SYSTEM

In the doppler system the transmitting-array elements are activated sequentially, giving the effect of a "virtual velocity" to each transmission. Consequently the signal received at a given point within the beam is shifted in wavelength by an amount proportional to $\sin \theta$, where θ is the angle of the point off the beam center. In the model described here, doppler transmissions are accommodated by specifying a "virtual velocity" vector $\mathbf{V}\mathbf{V}$ to be associated with each transmission. Then all field components *initially* reflected from a given reflector S_i have wavelength λ_i given by

$$\lambda_i = \lambda - \frac{\lambda}{c} \left[\frac{(\mathbf{V}\mathbf{C}_i - \mathbf{R}_T) \cdot \mathbf{V}\mathbf{V}}{|\mathbf{V}\mathbf{C}_i - \mathbf{R}_T|} \right],$$

where c is the speed of light.

Let $E_i(\mathbf{R}_P)$ be the component of the field at P due to all transmissions originating at \mathbf{R}_T initially scattered from reflector S_i (the field component at P with wavelength λ_i). Then

$$\begin{aligned} \mathbf{E}_i(\mathbf{R}_P) = \mathbf{G}_i(\mathbf{R}_P) + \sum_{\substack{j=1 \\ j \neq i}}^M \left[\mathbf{G}_{ij}(\mathbf{R}_P) + \sum_{\substack{k=1 \\ k \neq j}}^M \left[\mathbf{G}_{ijk}(\mathbf{R}_P) + \dots \right. \right. \\ \left. \left. + \sum_{\substack{n=1 \\ n \neq n-1}}^M \left[\mathbf{G}_{ijk \dots n} \dots \right] \dots \right] \right], \end{aligned} \quad (42)$$

and the total field at P as given by Eq. (41) becomes

$$\mathbf{E}_T(\mathbf{R}_P) = \mathbf{E}_O(\mathbf{R}_P) + \sum_{i=1}^M \mathbf{E}_i(\mathbf{R}_P), \quad (43)$$

where $|\mathbf{E}_i(\mathbf{R}_P)|$ represents the line in the energy spectrum at P corresponding to wavelength λ_i .

CONCLUSIONS

The model described herein calculates the standard far-field approximation to the Helmholtz integrals representing the field multiply scattered by flat polygonal metallic reflectors. The new representation of the far-field approximation to the Helmholtz integral as a finite sum, as developed in Ref. 2, has been employed to achieve a great economy in calculation. More generally, it is known that the Helmholtz integral has an exact closed-form representation as a line integral evaluated over the boundary of the reflecting surface. An expansion of the model is being developed to include the treatment of curved reflecting surfaces, along with the exact line-integral representations of the Helmholtz integral, applicable to plane-wave, spherical-wave, and dipole radiation. Also under investigation is the possibility of developing a closed-form expression for the Helmholtz integral which would involve no integrations at all, similar to that obtained and used for the far-field approximation to the exact Helmholtz integral.

REFERENCES

1. W.B. Gordon, "Far-Field Approximations to the Kirchhoff-Helmholtz Representations of Scattered Fields," to appear in *IEEE Trans. Antennas Propag.*, July, 1975.
2. W.B. Gordon, "Evaluating the Helmholtz Integral: Part I - Basic Theory," NRL Report 7792, Aug. 21, 1974.
3. S. Siver, editor, *Microwave Antenna Theory and Design*, Dover, New York, 1965, p. 173.

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C
50 CONTINUE
END

	IDENT	FIELD
PROGRAM LENGTH		00146
ENTRY POINTS	FIELD	00070
BLOCK NAMES		
	CONST	00003
	INPUT	01243
	PRELIM	06253
EXTERNAL SYMBOLS		
	QBQENTRY	
	QBQDICT.	
	READIN	
	SETUP	
	FPGEN	
	DIRECT	
	REFLEC	
	PRNT	
00053 SYMBOLS		

```

SUBROUTINE FPGEN(NP,RP, ID)
C
C THIS SUBROUTINE TO BE FURNISHED BY THE USER
C
C SUBROUTINE FPGEN GENERATES THE FIELD POINT RP AT WHICH THE DIRECT
C AND SCATTERED FIFLWS ARE TO BE EVALUATED
C
C PARAMETERS
C NP = FIELD POINT STATUS (NP = 0 FOR INITIAL ENTRY)
C RP = GENERATED FIELD POINT
C ID = DATA SET ID
C
COMMON/CONST/PI,TWOPI,CM
DIMENSION RP(3)
10 IF(NP .NE. 0) GO TO 20
C
C INITIALIZE GENERATOR
C
C THIS EXAMPLE WILL CALCULATE THE 10 SAMPLE POINTS
C RP(1) = 100.*COS(N*PI/180.)
C RP(2) = 0.
C RP(3) = 15.*SQRT(3.) + 100.*SIN(N*PI/180.)
C FOR N FROM 1 THROUGH 10 ON 10 SUCCESSIVE ENTRIES
C
NMAX = 10
CONST = 15.0*SQRT(3.0)
RP(2) = 0.
NP = 1
N = 0
20 N = N + 1
IF(N .GT. NMAX) GO TO 30
FN = N - 1
ARG = FN*PI/180.
RP(1) = 100.*COS(ARG)
RP(3) = CONST + 100.*SIN(ARG)
RETURN
30 NP = -1
RETURN
END

```

	IDENT	FPGEN
PROGRAM LENGTH		00136
ENTRY POINTS	FRGEN	00003
BLOCK NAMES		
	CONST	00003
EXTERNAL SYMBOLS		
	QBQDICT.	
	SQRTF	
	SINF	
	COSF	
00046 SYMBOLS		

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SUBROUTINE READIN

```

C
C SUBROUTINE READIN INPUTS AND PRINTS THE FOLLOWING DATA
C ITITLE = 80 CHARACTER RUN IDENTIFICATION
C XLMDA = TRANSMITTING WAVELENGTH (METERS)
C VV(3) = VIRTUAL VELOCITY (METERS/SEC)
C ET(3) = TRANSMITTED RADIATION
C RT(3) = TRANSMITTER POSITION COORDINATES
C D(3) = DIRECTION OF MAIN BEAM
C
C NBOUND = MAX. NO. OF REFLECTED BOUNCES ALLOWED
C MP = NUMBER OF REFLECTING PLATES
C NV(I) = NUMBER OF VERTICES OF PLATE I
C V(K,J,I) = COORDINATE K OF VERTEX J OF PLATE I
C
C COMMON/INPUT/ITITLE(10),XLMDA,VV(3),ET(3),RT(3),D(3),NBOUND,MP,
1 NV(50),V(3,4,50)
C INTEGER WT
C
C CURRENT DIMENSIONS ALLOW UP TO 30 REFLECTORS (PRELIM, PROJ) AND
C UP TO 4 (INPUT, PROJ) VERTICES PER REFLECTOR
C INPUT IS DIMENSIONED FOR 50 REFLECTORS
C
C ASSIGN INPUT, OUTPUT LOGICAL TAPE UNITS
C IN = 60
C WT = 61
C READ(IN,100) (ITITLE(I),I=1,10),XLMDA,(VV(I),I=1,3),(ET(I),I=1,3)
1 (RT(I),I=1,3), (D(I),I=1,3),NBOUND,MP
100 FORMAT(10A8/8F10.3/5F10.3/2I5)
C WRITE(WT,200) (ITITLE(I),I=1,10),XLMDA,(VV(I),I=1,3),(ET(I),I=1,3)
1 (RT(I),I=1,3), (D(I),I=1,3),NBOUND,MP
200 FORMAT(1H1/10A8, //5X,10HWAVELENGTH,16X,F10.3//5X,16HVIRTUAL VELOCIT
1Y,10X3F10.3//5X,51HTRANSMITTED RADIATION,5X,3F10.3//5X,23HTRANSMI
2TTER COORDINATES,3X,3F10.3//5X,19HMAIN BEAM DIRECTION,7X,3F10.3//
35X,16HMAX. NO. BOUNCES,10X,110//5X,21HNO. REFLECTING PLATES,5X,
4I10)
C
C READ(IN,110) (NV(I),I=1,MP)
110 FORMAT(16I5)
C DO 130 I = 1,MP
C N = NV(I)
C READ(IN,120) ((V(K,J,I),K=1,3),J=1,N)
120 FORMAT(8F10.3)
C WRITE(WT,210) I, N, ((V(K,J,I),K=1,3),J=1,N)
210 FORMAT(///5X,5HPLATE,I10,5X,12HNO. VERTICES,I10//1X,12F10.3)
130 CONTINUE
C
C RETURN
C END

```

	IDENT	READIN
PROGRAM LENGTH	00510	
ENTRY POINTS	READIN	00134
BLOCK NAMES	INPUT	01243
EXTERNAL SYMBOLS	THEND.	
	QBODICT.	
	TSH.	
	STH.	
	QNSINGL.	

00104 SYMBOLS

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```

SUBROUTINE SETUP
C
C SUBROUTINE SETUP MAKES PRELIMINARY GEOMETRICAL CALCULATIONS
C (COMMON PRELIM) DEPENDENT UPON INPUT REFLECTING PLATE DATA
C AND ALSO COMPUTES THE DOPPLER WAVELENGTH AND WAVENUMBER
C ASSOCIATED WITH EACH REFLECTING PLATE
C
COMMON/CONST/PI,TWOPI,CM
COMMON/INPUT/ITITLE(10),XLMDA,VV(3),ET(3),RT(3),D(3),NBOUND,MP,
1 NV(50),V(3,4,50)
COMMON/PRELIM/M,WAVFNM,VC(3,50),VNORM(3,50),VA(50),VASUM,XLAM(50),
1 XK(50),R(3,31,30)
DIMENSION CS(3),V1V(3),V2V(3)
C
C CURRENT DIMENSIONS ALLOW UP TO 30 REFLECTORS (PRELIM, PROJ) AND
C UP TO 4 (INPUT, PROJ) VERTICES PER REFLECTOR
C INPUT IS DIMENSIONED FOR 50 REFLECTORS
C
C SET NO. OF EFFECTIVE PLATES EQUAL TO NO. OF PLATES INPUT
M = MP
C
C CALCULATE CENTROID VC OF EACH PLATE
DO 35 J = 1,M
  CS(1) = 0.
  CS(2) = 0.
  CS(3) = 0.
  N = NV(J)
  DO 25 I = 1,N
    DO 20 K = 1,3
      CS(K) = CS(K) + V(K,I,J)
    20 CONTINUE
  25 CONTINUE
  FNV = FLOAT(N)
  DO 30 K = 1,3
    VC(K,J) = CS(K)/FNV
  30 CONTINUE
  35 CONTINUE
C
C CALCULATE UNIT NORMAL VNORM TO EACH PLATE
DO 70 J = 1,M
  VS1 = 0.
  VS2 = 0.
  DO 55 K = 1,3
    V1V(K) = V(K,1,J) - VC(K,J)
    V2V(K) = V(K,2,J) - VC(K,J)
    VS1 = VS1 + V1V(K)*V1V(K)
    VS2 = VS2 + V2V(K)*V2V(K)
  55 CONTINUE
  VM = SQRT(VS1)*SQRT(VS2)
  VNORM(1,J) = (V1V(2)*V2V(3) - V1V(3)*V2V(2))/VM
  VNORM(2,J) = (V1V(3)*V2V(1) - V1V(1)*V2V(3))/VM
  VNORM(3,J) = (V1V(1)*V2V(2) - V1V(2)*V2V(1))/VM
C CHECK SIGN AND NORMALIZE
  VNM = SQRT(VNORM(1,J)*VNORM(1,J) + VNORM(2,J)*VNORM(2,J) +

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```

1      VNORM(3,J)*VNORM(3,J))
      DOT = 0.
      DO 60 K = 1,3
        DOT = DOT + VNORM(K,J)*(VC(K,J) - RT(K))
60     CONTINUE
      IF(DOT .GT. 0.) VNM = -VNM
      DO 65 K = 1,3
        VNORM(K,J) = VNORM(K,J)/VNM
65     CONTINUE
70    CONTINUE

C
C
C      CALCULATE AREA VA OF EACH PLATE
      VASUM = 0.
      DO 85 J = 1,M
        AREA = 0.
        N = NV(J)
        DO 80 I = 1,N
          A = 0.
          B = 0.
          C = 0.
C      COMPUTE SIDES OF TRIANGLE J
          II = I + 1
          IF (II .GT. N) II = 1
          DO 75 K = 1,3
            A = A + (V(K,I,J) - VC(K,J))**2
            B = B + (V(K,II,J) - VC(K,J))**2
            C = C + (V(K,I,J) - V(K,II,J))**2
75     CONTINUE
C      CALCULATE AREA OF TRIANGLE J
          DELA = SQRT(2.0*(A*B + B*C + C*A) - (A*A + B*B + C*C))/4.0
          AREA = AREA + DELA
80     CONTINUE
        VA(J) = AREA
        VASUM = VASUM + AREA
85    CONTINUE

C
C
C      CALCULATE PLATE (I-1) TO PLATE (J) UNIT DIRECTION VECTORS R(K,I,J)
C      WITH RESPECT TO PLATE CENTROIDS VC
C      NOTE 1.LE.I.LE.M+1 SO THAT I REFERS TO PLATE I-1, WHERE PLATE 0
C      IS THE TRANSMITTER
C
C      CALCULATE TRANSMITTER (PLATE 0) TO PLATE J UNIT DIRECTION VECTORS
      DO 115 J = 1,M
        VM = 0.
        DO 105 K = 1,3
          V1V(K) = VC(K,J) - RT(K)
          VM = VM + V1V(K)*V1V(K)
105     CONTINUE
        VM = SQRT(VM)
        DO 110 K = 1,3
          R(K,1,J) = V1V(K)/VM
110     CONTINUE
115    CONTINUE
C

```

```

C ..... CALCULATE REMAINING DIRECTION VECTORS .....
DO 135 I = 1,M
  II = I + 1
  DO 130 J = 1,M
    IF(J .EQ. I) GO TO 130
    VM = 0.
    DO 120 K = 1,3
      V1V(K) = VC(K,J) - VC(K,I)
      VM = VM + V1V(K)*V1V(K)
    120 CONTINUE
    VM = SQRT(VM)
    DO 125 K = 1,3
      R(K,II,J) = V1V(K)/VM
    125 CONTINUE
  130 CONTINUE
135 CONTINUE

C
C
C
C CHECK FOR DOPPLER
VVM = 0.
DO 138 K = 1,3
  VVM = VVM + VV(K)*VV(K)
138 CONTINUE
VVM = SQRT(VVM)
IF(VVM .EQ. 0.0) GO TO 150
C CALCULATE DOPPLER WAVELENGTH XLAM AND WAVENUMBER XK AS SEEN
C BY EACH PLATE
DO 145 J = 1,M
  VS1 = 0.
  VS2 = 0.
  DO 140 K = 1,3
    CS(K) = VC(K,J) - RT(K)
    VS1 = VS1 + CS(K)*CS(K)
    VS2 = VS2 + VV(K)*CS(K)
  140 CONTINUE
  VS1 = SQRT(VS1)
  XLAM(J) = XLMDA*(1.0 + VS2/(CM*VS1))
  XK(J) = TWOPI/XLAM(J)
145 CONTINUE
GO TO 160

C
C NO DOPPLER
C
C CALCULATE WAVE NUMBER
150 WAVENM = TWOPI/XLMDA
DO 155 J = 1,M
  XLAM(J) = XLMDA
  XK(J) = WAVENM
155 CONTINUE

C
160 RETURN
END

```

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	IDENT	SETUP
PROGRAM LENGTH		00740
ENTRY POINTS	SETUP	00014
BLOCK NAMES		
	CONST	00003
	INPUT	01243
	PRELIM	06253
EXTERNAL SYMBOLS		
	QBQDICT.	
	SQRTE	
00203 SYMBOLS		

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```

SUBROUTINE DIRECT(RP)
C
C SUBROUTINE DIRECT CALCULATES THE DIRECT FIELD COMPONENT ED FROM
C TRANSMITTER RT TO FIELD POINT RP
C
COMMON/CONST/PI,TWOPI,CM
COMMON/INPUT/ITITLE(10),XLMDA,VV(3),ET(3),RT(3),D(3),NBOUND,MP,
1 NV(50),V(3,4,50)
COMMON/PRELIM/M,WAVENM,VC(3,50),VNORM(3,50),VA(50),VASUM,XLAM(50),
1 XK(50),R(3,31,30)
COMMON/FIELD/ED(3),ES(3,50)
COMPLEX ED,ES
COMPLEX FAC
DIMENSION RP(3),WS(3)
C
C CURRENT DIMENSIONS ALLOW UP TO 30 REFLECTORS (PRELIM, PROJ) AND
C UP TO 4 (INPUT, PROJ) VERTICES PER REFLECTOR
C INPUT IS DIMENSIONED FOR 50 REFLECTORS
C
C DIPOLE
DOT = 0.
DO 10 K = 1,3
WS(K) = RP(K) - RT(K)
DOT = DOT + ET(K)*WS(K)
10 CONTINUE
WM2 = WS(1)*WS(1) + WS(2)*WS(2) + WS(3)*WS(3)
WM = SQRT(WM2)
DW = DOT/WM2
C CALCULATE WAVE NUMBER (INCLUDING DOPPLER IF PRESENT)
DO 12 K = 1,3
DOT = DOT + WS(K)*VV(K)
12 CONTINUE
FLAM = XLMDA*(1.0 - DOT/(CM*WM))
WAVENM = TWOPI/FLAM
WK = WAVENM*WM
FAC = CMPLX(COS(WK),SIN(WK))
DO 15 K = 1,3
ED(K) = (ET(K) - DW*WS(K))*FAC/WM
15 CONTINUE
C PRINT
CALL PRNT(RP,RP,-1)
C RETURN
C
END

```

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	IDENT	DIRECT
PROGRAM LENGTH	00205	
ENTRY POINTS	DIRECT	00006
BLOCK NAMES		
	CONST	00003
	INPUT	01243
	PRELIM	06253
	FIELD	00462
EXTERNAL SYMBOLS		
	Q1004330	
	Q1005310	
	QBQDICT.	
	PRNT	
	SortF	
	SINF	
	COSF	
00076 SYMBOLS		

```

SUBROUTINE REFLEC(RP,PRTL,NBMAX)
C
C SUBROUTINE REFLEC CALCULATES THE REFLECTED (SCATTERED) FIELD
C COMPONENT ES(K,J) FOR EACH WAVELENGTH XLAM(J) AT POINT RP
C
COMMON/CONST/PI,TWOPI,CM
COMMON/INPUT/ITITLE(10),XLMDA,VV(3),ET(3),RT(3),D(3),NBOUND,MP,
1 NV(50),V(3,4,50)
COMMON/PRELIM/M,WAVFNM,VC(3,50),VNORM(3,50),VA(50),VASUM,XLAM(50),
1 XK(50),R(3,31,30)
COMMON/FIELD/ED(3),FS(3,50)
COMPLEX ED,ES,CW
DIMENSION VJT(3),FOJ(3),PRTL(50)
DIMENSION NB(50)
COMPLEX F,G
DIMENSION NP(10),F(3,10),G(3)
DIMENSION RR(3)
C
C CURRENT DIMENSIONS ALLOW UP TO 30 REFLECTORS (PRELIM) AND
C UP TO 4 (INPUT) VERTICES PER REFLECTOR
C INPUT IS DIMENSIONED FOR 50 REFLECTORS
C
C INITIALIZE SCATTERED FIELD COMPONENTS TO ZERO
DO 15 J = 1,M
DO 10 K = 1,3
ES(K,J) = 0.
10 CONTINUE
15 CONTINUE
C
C INITIALIZE BOUNCE COUNTERS
DO 20 J = 1,M
NB(J) = 0
20 CONTINUE
C
C SET EPSILON FOR FIELD STOPPING RULE
EPS = 0.01
C CALCULATE FACTOR FOR FIELD STOPPING RULE AT GIVEN PLATE
FAC = EPS*CVMAG(ED)/VASUM
C
C *****
C
DO 200 J = 1,M
C
C CALCULATE DIRECT FIELD AT PLATE J (SCAT PROVIDES 1/R FACTOR)
C DIPOLE
DOT = 0.
DO 30 K = 1,3
VJT(K) = VC(K,J) - RT(K)
DOT = DOT + ET(K)*VJT(K)
30 CONTINUE
VJTM2 = VJT(1)*VJT(1) + VJT(2)*VJT(2) + VJT(3)*VJT(3)
VJTM = SQRT(VJTM2)
DV = DOT/VJTM2
DO 33 K = 1,3
FOJ(K) = (ET(K) - DV*VJT(K))

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33 CONTINUE
C
C CHECK DIRECT FIELD AGAINST STOPPING RULE
  FMIN = VA(J)*FAC
  FMAG = SQRT(FOJ(1)*FOJ(1) + FOJ(2)*FOJ(2) + FOJ(3)*FOJ(3))
  IF (FMAG .LT. FMIN) GO TO 200
C CHECK BOUNCE COUNTER AGAINST MAXIMUM ALLOWED
  IF (NB(J) .GE. NBOUND) GO TO 200
C
C INCREMENT BOUNCE COUNTER
  NR(J) = NB(J) + 1
C
C REFLECT FOJ WITH RESPECT TO PLATE J
  DOT = 0.
  DO 45 K = 1,3
    DOT = DOT + FOJ(K)*VNORM(K,J)
45 CONTINUE
  DO 48 K = 1,3
    F(K,1) = 2.0*DOT*VNORM(K,J) - FOJ(K)
48 CONTINUE
C
C COMPUTE FIELD COMPONENT AT RP DUE TO SCATTER OF F(1) FROM PLATE J
  CALL SCAT(0,J,XK(J),F(1,1),RP,G)
C
C ADD TO SCATTER FIELD COMPONENT FOR XLAM(J)
  DO 55 K = 1,3
    ES(K,J) = ES(K,J) + G(K)
55 CONTINUE
C
C *****
C
C CALCULATE TOTAL SCATTERED FIELD COMPONENT AT RP WITH WAVELENGTH
  XLAM(J)
C NB(J) = MAX. NO. BOUNCES INVOLVED IN COMPUTATION OF FIELD
  COMPONENT WITH WAVELENGTH XLAM(J)
C NP(N) = CURRENT PLATE NO. FOR NTH BOUNCE
C N = CURRENT BOUNCE NO.
C
  NP(1) = J
  DO 60 N = 2,NBOUND
    NP(N) = 1
60 CONTINUE
C INITIALIZE BOUNCE SEQUENCE
  70 N = 2
  75 NP(N) = NP(N) + 1
  IF (NP(N) .GT. M) GO TO 95
C CHECK FOR DISTINCT BOUNCE
  78 IF (NP(N) .EQ. NP(N-1)) GO TO 75
C COMPUTE F(N), THE VALUE OF FIELD F(N-1) SCATTERED BY PLATE NP(N-1)
  AT THE POINT VC(NP(N))
  NP1 = 0
  IF (N .GT. 2) NP1 = NP(N-2)
  NP2 = NP(N-1)
  CALL SCAT(NP1,NP2,XK(J),F(1,N-1),VC(1,NP(N)),F(1,N))
  FMIN = VA(NP2)*FAC
  FMAG = CVMAG(F(1,N))

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      IF (FMAG .LT. FMIN) GO TO 75
C
C   REFLECT F WITH RESPECT TO PLATE NP2
80  DO 82 K = 1,3
      DOT = DOT + F(K,N)*VNORM(K,NP2)
82  CONTINUE
      DO 85 K = 1,3
          F(K,N) = 2.0*DOT*VNORM(K,NP2) - F(K,N)
85  CONTINUE
C
C   CALCULATE FIELD COMPONENT AT RP DUE TO SCATTER OF F(N) FROM
C   PLATE NP(N)
      CALL SCAT(NP2,NP(N),XK(J),F(1,N),RP,G)
C
C   ADD TO CURRENT SCATTERED FIELD COMPONENT
      DO 90 K = 1,3
          ES(K,J) = ES(K,J) + G(K)
90  CONTINUE
C
C   UPDATE MAX. NO. OF BOUNCES FOR THIS WAVELENGTH, IF APPROPRIATE
      IF (N .GT. NB(J)) NB(J) = N
C
C   STEP UP BOUNCE LEVEL
      N = N + 1
      IF (N .LE. NBOUND) GO TO 75
C
C   END OF BOUNCE STRING
C   STEP BACK ONE BOUNCE LEVEL AND INCREMENT
95  N = N - 1
      IF (N .GT. 1) GO TO 75
C
C
C   FINISHED WITH WAVELENGTH XLAM(J) CONTRIBUTION TO FIELD
200 CONTINUE
C
C   *****
C
C   SET UP PRINT FREQUENCY ARRAY
      DO 205 J = 1,M
          PRTL(J) = CM/XLAM(J)
205 CONTINUE
C
C   CHECK FOR DISTINCT SCATTER FREQUENCIES
      MM = M - 1
      DO 230 I = 1,MM
          II = I + 1
          IF (PRTL(I) .EQ. -1.0) GO TO 230
          DO 225 J = II,M
              WS = ABSF(PRTL(I) - PRTL(J))
              IF (WS .GT. 1000.) GO TO 225
              DO 220 K = 1,3
                  ES(K,I) = FS(K,I) + ES(K,J)
220          CONTINUE
              PRTL(J) = -1.0
225          CONTINUE

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230 CONTINUE
C
C   DETERMINE MAXIMUM NUMBER OF BOUNCES
   NBMAX = 0
   DO 235 J = 1,M
     IF(NB(J) .GT. NBMAX) NBMAX = NB(J)
235 CONTINUE
C
   RETURN
   END
```

	IDENT	REFLEC
PROGRAM LENGTH	01157	
ENTRY POINTS	REFLEC	00207
BLOCK NAMES		
	CONST	00003
	INPUT	01243
	PRELIM	06253
	FIELD	00462
EXTERNAL SYMBOLS		
	Q1Q02330	
	Q1Q04310	
	Q1Q02310	
	Q1Q03330	
	QBQDICT.	
	CVMAG	
	SCAT	
	SQRTF	
00257 SYMBOLS		

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SUBROUTINE SCAT(I,J,FK,F,RP,G)
C
C SUBROUTINE SCAT CALCULATES THE VALUE OF THE SCATTERED FIELD (FIELD
C F SCATTERED FROM PLATE J) AT POINT R
C
C PARAMETERS
C I = SOURCE PLATE NUMBER
C J = REFLECTOR PLATE NUMBER
C FK = WAVE NUMBER
C F = REFLECTED FIELD AT CENTROID OF PLATE J WHOSE SOURCE IS
C THE CENTROID OF PLATE I (COMPLEX)
C RP = POINT AT WHICH SCATTERED FIELD IS TO BE EVALUATED
C G = FIELD AT POINT R ARISING FROM THE SCATTER OF FIELD F
C FROM PLATE J
C
COMMON/CONST/PI,TWOPI,CM
COMMON/INPUT/ITITLE(10),XLMDA,VV(3),ET(3),RT(3),D(3),NBOUND,MP,
1 NV(50),V(3,4,50)
COMMON/PRELIM/M,WAVENM,VC(3,50),VNORM(3,50),VA(50),VASUM,XLAM(50),
1 XK(50),R(3,31,30)
DIMENSION VIJ(3),W(3),WN(3)
DIMENSION F(3),FP(3),RP(3),G(3),RV(3),DELV(3)
COMPLEX F,FP,G,CF,XFACT,T,TN
C
C CURRENT DIMENSIONS ALLOW UP TO 30 REFLECTORS (PRELIM) AND
C UP TO 4 (INPUT) VERTICES PER REFLECTOR
C INPUT IS DIMENSIONED FOR 50 REFLECTORS
C
Eps = 0.001
C
C
C CALCULATE FIELD WITH PHASE SHIFT
C CHECK FOR TRANSMITTER (I = 0)
IF( I .EQ. 0) GO TO 3
DO 2 K = 1,3
VIJ(K) = VC(K,J) - VC(K,I)
2 CONTINUE
GO TO 5
3 DO 4 K = 1,3
VIJ(K) = VC(K,J) - RT(K)
4 CONTINUE
5 VIJM = SQRT( VIJ(1)*VIJ(1) + VIJ(2)*VIJ(2) + VIJ(3)*VIJ(3) )
ARG = FK*VIJM
CF = CMPLX(COS(ARG)/VIJM,SIN(ARG)/VIJM)
C
DO 6 K = 1,3
FP(K) = CF*F(K)
6 CONTINUE
C
C CHECK IF FIELD SOURCE AND FIELD RECEIVER RP ARE ON SAME SIDE OF
C PLATE J
DO 7 K = 1,3
RV(K) = RP(K) - VC(K,J)
7 CONTINUE

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RVM = SQRT(RV(1)*RV(1) + RV(2)*RV(2) + RV(3)*RV(3))
DOT = 0.
C
DRVN = 0.
DO 10 K = 1,3
  DRVN = DRVN + RV(K)*VNORM(K,J)
10 CONTINUE
IF(DRVN .GT. 0.0) GO TO 25
DO 20 K = 1,3
  G(K) = (0.,0.)
20 CONTINUE
GO TO 95
C
C   NORMALIZE VIJ AND RV VECTORS
C   INITIATE W VECTOR
25 DO 27 K = 1,3
  VIJ(K) = VIJ(K)/VIJM
  RV(K) = RV(K)/RVM
  W(K) = VIJ(K) - RV(K)
27 CONTINUE
C
C
DOT = 0.
DO 28 K = 1,3
  DOT = DOT + W(K)*VNORM(K,J)
28 CONTINUE
C
W2 = 0.
DO 30 K = 1,3
  W(K) = W(K) + DOT*VNORM(K,J)
  WM2 = WM2 + W(K)*W(K)
30 CONTINUE
C
C   CHECK FOR SPECULAR REFLECTION
C
IF (WM2 .GT. 0.0) GO TO 45
C
C   SPECULAR REFLECTION IN DIRECTION OF RP
WS = (FK*DRVN*VA(J))/TWOPI
ARG = FK*RVM
CF = CMPLX(WS*SIN(ARG),-WS*COS(ARG))
C
DO 40 K = 1,3
  G(K) = CF*FP(K)
40 CONTINUE
GO TO 95
C
C   BEGIN EQ. 4.6
45 DOT = 0.
DO 50 K = 1,3
  DOT = DOT + (RV(K) - VIJ(K))*VNORM(K,J)
50 CONTINUE
C
WS = DOT/(WM2*2.0*TWOPI*RVM)
ARG = FK*RVM
XFACT = CMPLX(WS*COS(ARG),WS*SIN(ARG))

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C
  T = (0.,0.)
  NN = NV(J)
  DO 55 N = 1,NN
    N1 = N + 1
    IF(N1 .GT. NN) N1 = 1
    DO 55 K = 1,3
      DELV(K) = V(K,N1,J) - V(K,N,J)
55  CONTINUE
C
  DOT = 0.
  DO 60 K = 1,3
    DOT = DOT + W(K)*DELV(K)
60  CONTINUE
  IF(DOT .NE. 0.0) GO TO 65
  S = 1.0
  GO TO 70
C
65  WS = FK*DOT
  S = SIN(WS)/WS
C
C
70  WN = (W)*X(VNORM)
  WN(1) = W(2)*VNORM(3,J) - W(3)*VNORM(2,J)
  WN(2) = W(3)*VNORM(1,J) - W(1)*VNORM(3,J)
  WN(3) = W(1)*VNORM(2,J) - W(2)*VNORM(1,J)
C
  DOT = 0.
  DO 75 K = 1,3
    DOT = DOT + WN(K)*DELV(K)
75  CONTINUE
  WS = S*DOT
C
C
  DOT = 0.
  DO 80 K = 1,3
    DOT = DOT + W(K)*(V(K,N,J) + V(K,N1,J))
80  CONTINUE
  ARG = FK*DOT/2.
C
  TN = CMPLX(WS*COS(ARG),WS*SIN(ARG))
  T = T + TN
C
85 CONTINUE
C
  SCATTERED FIELD
  DO 90 K = 1,3
    G(K) = T*XFACT*FP(K)
90 CONTINUE
95 RETURN
C
  END

```

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	IDENT	SCAT
PROGRAM LENGTH	00765	
ENTRY POINTS	SCAT	00030
BLOCK NAMES		
	CONST	00003
	INPUT	01243
	PRELIM	06253
EXTERNAL SYMBOLS		
	Q1004330	
	Q1002330	
	QBQDICT.	
	SQRTF	
	SINF	
	COSF	
00232 SYMBOLS		

```

FUNCTION CVMAG(A)
C
C FUNCTION CVMAG COMPUTES THE MAGNITUDE OF THE COMPLEX
C THREE-DIMENSIONAL VECTOR A
C
C COMPLEX A,SUM
C DIMENSION A(3)
C
C SUM = 0.
Dn 10 I = 1,3
SUM = SUM + A(I)*CONJG(A(I))
10 CONTINUE
CVMAG = SQRT(SUM)
RETURN
END
    
```

	IDENT	CVMAG
PROGRAM LENGTH	00111	
ENTRY POINTS	CVMAG	00003
EXTERNAL SYMBOLS	Q1Q04330	
	Q1Q02330	
	Q8QDICT	
	SQRTF	
00031 SYMBOLS		

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```

SUBROUTINE PRNT(RP,PRTL,KEY)
C
C SUBROUTINE PRNT PRINTS THE DIRECT AND SCATTERED FIELD SPECTRUM
C EVALUATED AT POINT RP
C
COMMON/CONST/PI,TWOPI,CM
COMMON/INPUT/ITITLE(10),XLMDA,VV(3),ET(3),RT(3),D(3),NBOUND,MP,
1 NV(50),V(3,4,50)
COMMON/PRELIM/M,WAVERN,VC(3,50),VNORM(3,50),VA(50),VASUM,XLAM(50),
1 XK(50),R(3,31,30)
COMMON/FIELD/ED(3),FS(3,50)
COMPLEX ED,ES
INTEGER WT
DIMENSION RR(1),PRTL(1)
COMMON/FLAG/IFLAG
C
C CURRENT DIMENSIONS ALLOW UP TO 30 REFLECTORS (PRELIM, PROJ) AND
C UP TO 4 (INPUT, PROJ) VERTICES PER REFLECTOR
C INPUT IS DIMENSIONED FOR 50 REFLECTORS
C
WT = 61
C CHECK FOR INITIAL ENTRY
IF(KEY .GE. 0) GO TO 10
C WRITE HEADING
GO TO (5,6), IFLAG
5 EDP = CABS(ED(1F))**2
GO TO 7
6 CONTINUE
C Y COMPONENT POWER
EDP = CABS(ED(2F))**2
7 CONTINUE
XFREQ = WAVENM*CM/TWOPI
WRITE (WT,110) (RP(K),K=1,3),XFREQ,EDP
110 FORMAT(//8X,11HFIELD POINT,5X,4HX = ,E11.3,3X,4HY = ,E11.3,3X,
1 4HZ = ,E11.3////21X,12HDIRECT FIELD/13X,10HFREQUENCY ,10X,
2 5HPOWER// 8X,E15.7,8X,E10.2 //19X,15HSCATTERED FIELD/13X,
3 10HFREQUENCY ,10X,5HPOWER,15X,3HN/S,15X,7HLOG N/S//)
LINECT = 5
RETURN
C
10 CONTINUE
GO TO (15,20), IFLAG
15 DO 18 I = 1,M
IF(PRTL(I) .EQ. -1.) GO TO 18
ESP = CABS(ES(1,I))**2
RSD = ESP/EDP
RSDL = ALOG10(RSD)
WRITE (WT,120) PRTL(I),ESP,RSD,RSDL
LINECT = LINECT + 1
IF(LINECT .LT. 60) GO TO 18
WRITE(WT,130)
LINECT = 1
18 CONTINUE
RETURN
20 DO 30 I = 1,M

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-----
IF(PRTL(I) .EQ. -1.) GO TO 30
ESP = CABS(ES(2,I))**2
RSD = ESP/EDP
RSDL = ALOG10(RSN)
-----
WRITE (WT,120) PRTL(I), ESP, RSD, RSDL
120 FORMAT(BX,E15.7,7X,E10.2,9X,E10.3,10X,E10.3/)
LINECT = LINECT + 1
IF(LINECT .LT. 60) GO TO 30
WRITE(WT,130)
LINECT = 1
130 FORMAT(1H1/)
30 CONTINUE
-----
C
C
RETURN
END
-----

```

	IDENT	PRNT
PROGRAM LENGTH	00451	
ENTRY POINTS	PRNT	00114
BLOCK NAMES		
	CONST	00003
	INPUT	01243
	PRELIM	06253
	FIELD	00462
	FLAG	00001
EXTERNAL SYMBOLS		
	THEND.	
	QBDDICT.	
	ALOG10	
	CABS	
	STH.	
	QNSINGL.	
00145 SYMBOLS		
LOAD		
RUN,2,3500		